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(54) **RESETTING DRIVE TRANSISTORS IN ELECTRONIC DISPLAYS**

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(57) **ABSTRACT**

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A method for resetting drive transistors associated with subpixels in an electroluminescent display, comprising providing an electroluminescent display having a plurality of subpixels, each subpixel including an electroluminescent device and a drive circuit having a drive transistor for providing current through its associated electroluminescent device; providing a separate aging signal for each subpixel during operation of the electroluminescent display after a predetermined operating time period by responding as a function of the current passing through each of the subpixels or as a function of a voltage associated with each drive circuit; comparing each of the separate aging signals with a corresponding threshold level to produce a separate staleness signal for each subpixel representing whether or not the associated drive transistor should be reset; and resetting the associated drive transistors in response to staleness signals that indicate such drive transistors should be reset.

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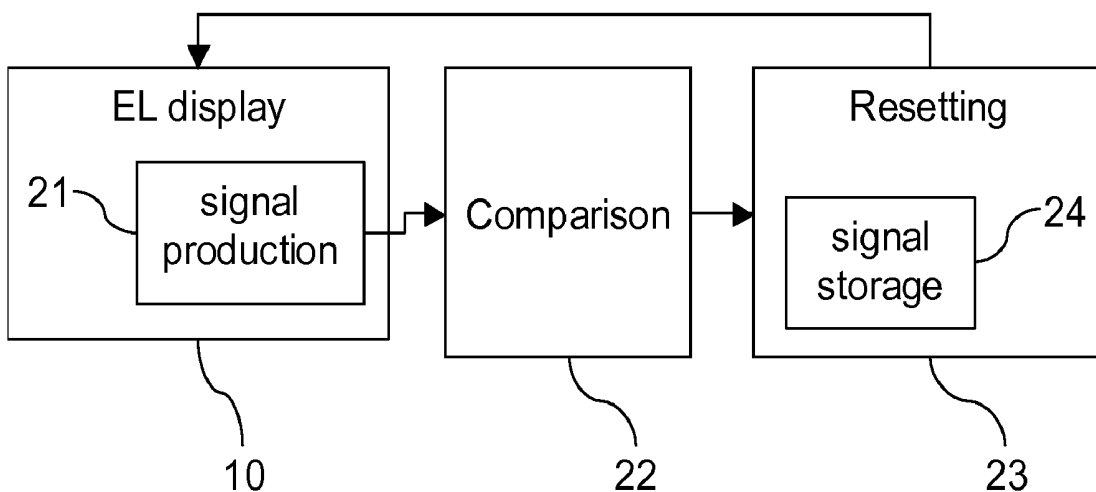


FIG. 1 (prior art)

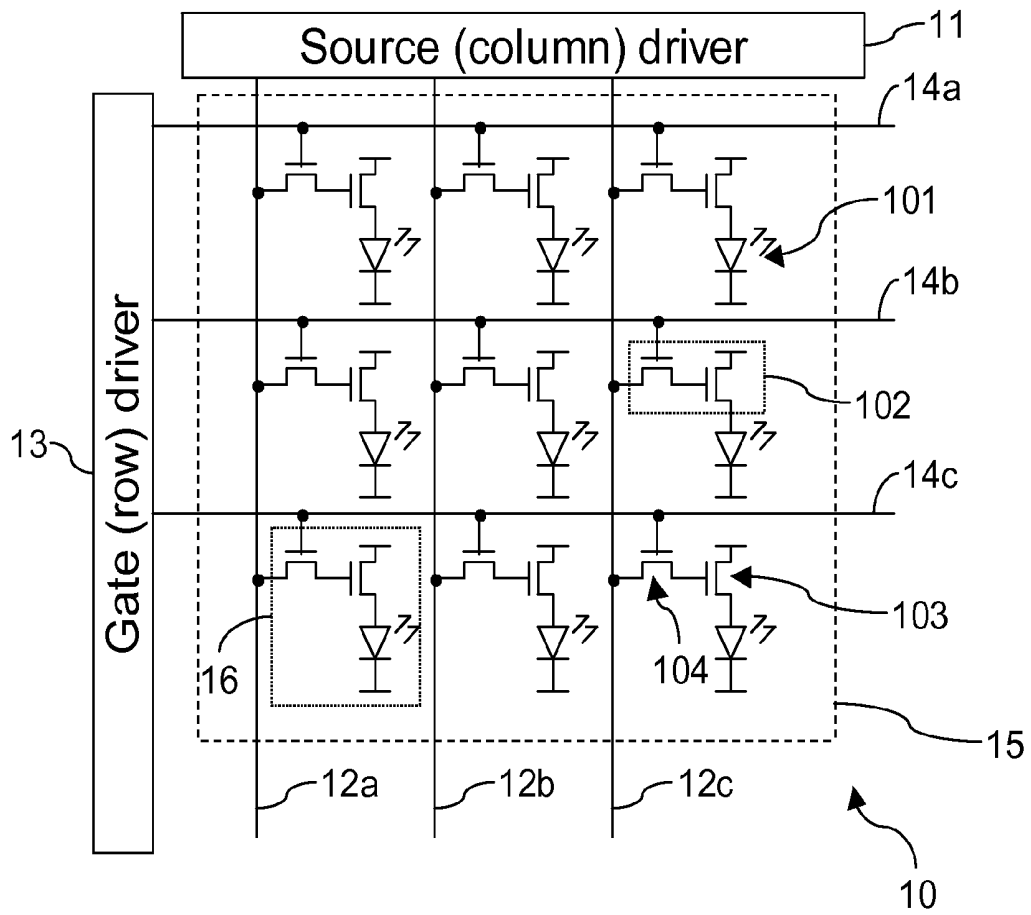


FIG. 2

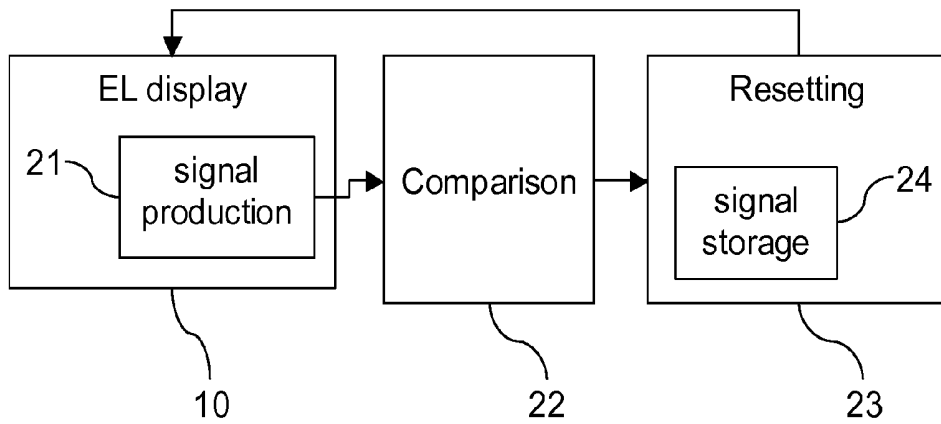


FIG. 3

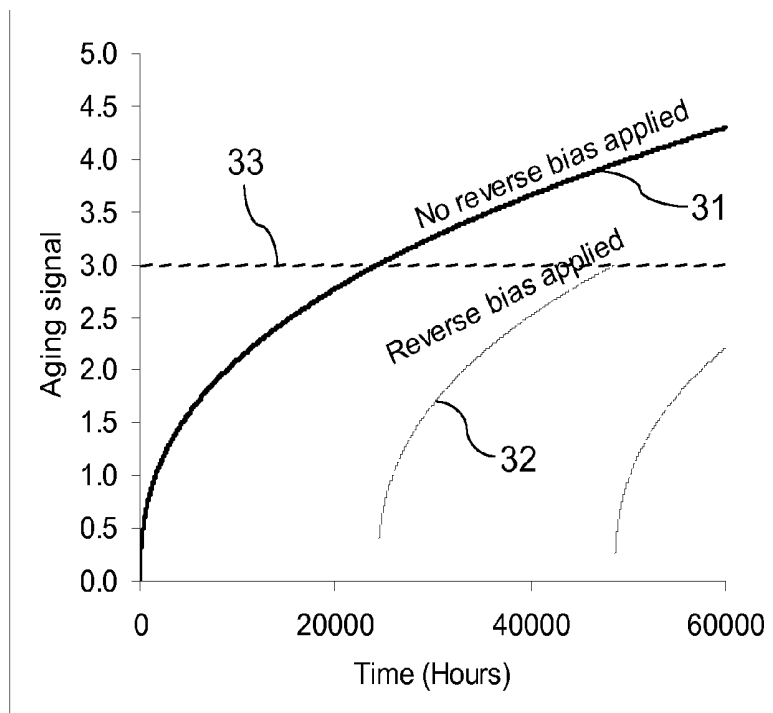


FIG. 4 (prior art)

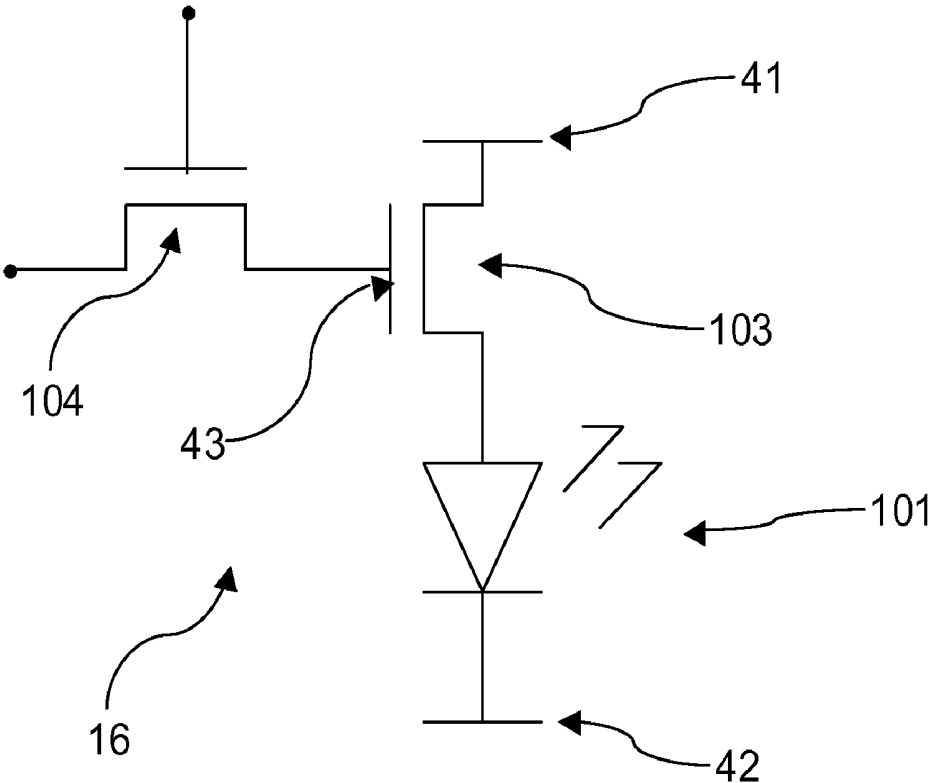
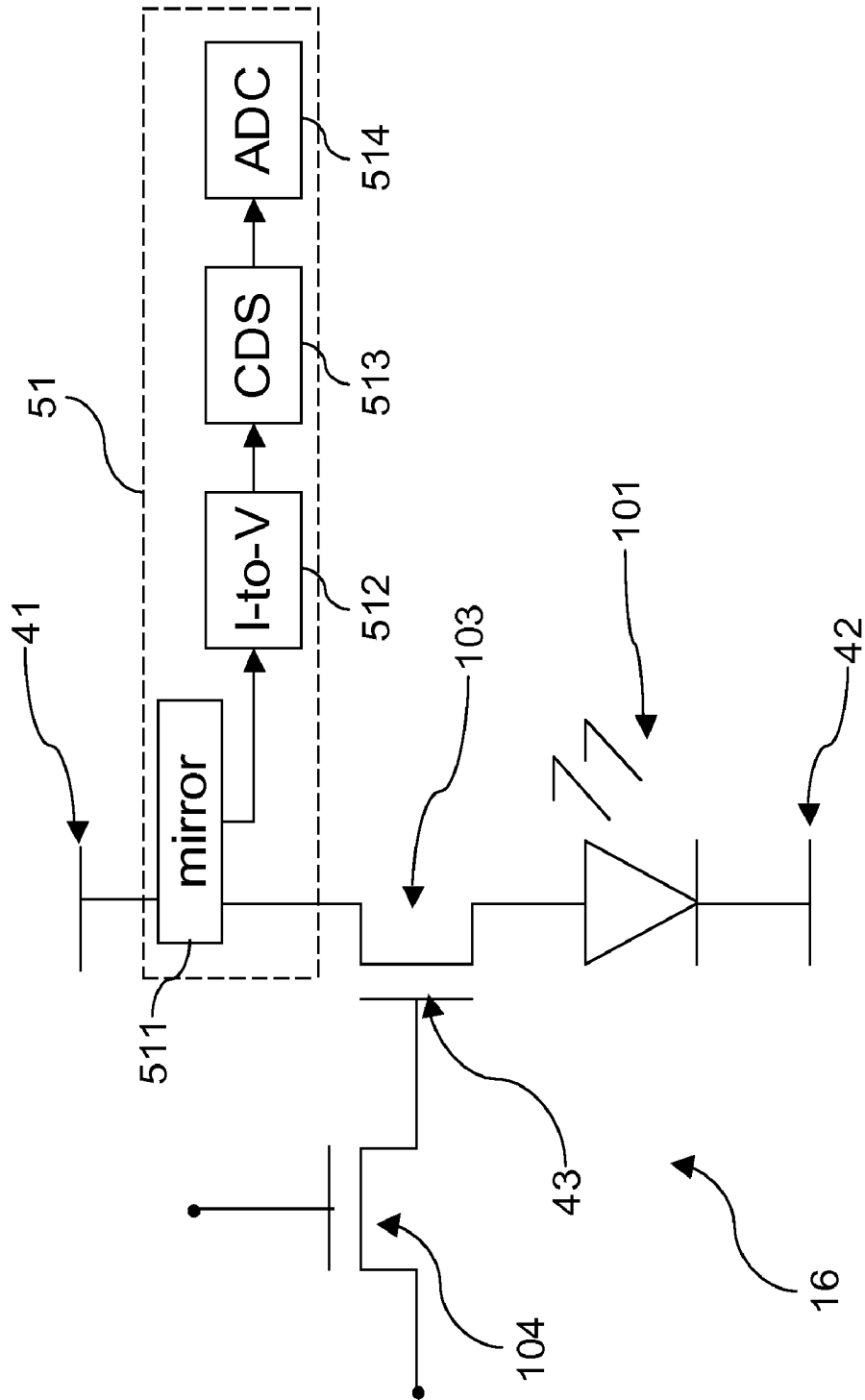


FIG. 5





## RESETTING DRIVE TRANSISTORS IN ELECTRONIC DISPLAYS

### CROSS-REFERENCE TO RELATED APPLICATIONS

[0001] Reference is made to commonly-assigned U.S. patent application Ser. No. 11/962,182, filed Dec. 21, 2007, entitled "Electroluminescent Display Compensated Analog Transistor Drive Signal" to Leon et al.; and commonly assigned U.S. patent application Ser. No. 11/766,823, filed Jun. 22, 2007, entitled "OLED Display With Aging and Efficiency Compensation" to Levey et al., the disclosures of which are herein incorporated by reference.

### FIELD OF THE INVENTION

[0002] The present invention relates to solid-state electroluminescent displays and more particularly to resetting drive transistors in such displays.

### BACKGROUND OF THE INVENTION

[0003] Solid-state electroluminescent (EL) displays are of great interest as an improved flat-panel display technology. These displays use current passing through thin films of material to generate light. Organic light-emitting diode (OLED) displays are a particularly promising technology employing thin films of organic material to generate the light. The color of light emitted and the efficiency of the energy conversion from current to light are determined by the composition of the organic thin-film material. Different organic materials emit different colors of light. A display can be formed as an array of pixels, each of which comprises one or more subpixels. For a color display, each subpixel can emit a different color of light.

[0004] In active-matrix OLED (AMOLED) and other active-matrix electroluminescent displays, current is typically supplied to the organic materials by drive transistors; these are generally thin-film transistors (TFTs). These TFTs are frequently made of amorphous silicon (a-Si), for example, as taught by Tanaka et al. in U.S. Pat. No. 5,034,340. Amorphous silicon is inexpensive and easy to manufacture. However, it is metastable: over time, as voltage bias is applied to the gate of an a-Si TFT, its threshold voltage ( $V_{th}$ ) shifts, thus shifting its I-V curve (Kagan & Andry, ed. *Thin-film Transistors*. New York: Marcel Dekker, 2003; Sec. 3.5, pp. 121-131).  $V_{th}$  typically increases over time under forward bias, so over time,  $V_{th}$  shift will, on average, cause a display to dim. This reduces the lifetime of the display. In addition, since the rate of  $V_{th}$  shift depends on applied bias, each individual subpixel can age at a rate different from other subpixels, resulting in display nonuniformity and visible image stick. This is a significant effect; most of the luminance loss of modern a-Si AMOLED displays is a result of changes in the amorphous silicon TFT performance rather than changes in the OLED.

[0005] The lack of stability in a-Si TFTs has been studied. For example, in an article entitled "Stability issues in digital circuits in amorphous silicon technology" published in *Electrical and Computer Engineering*, 2001, Vol. 1, pp. 583-588 by Mohan et al., the article discusses the fact that the  $V_{th}$  of an a-Si TFT can shift by as much as 2V when driven with a +20V bias for even 600 hours. This type of positive bias drive voltage is common for driving an OLED and this large threshold voltage shift can have a dramatic influence on the light output of the display. This same paper discusses the fact that

negative bias can have the opposite effect and, more importantly, that by cycling between a positive and negative bias, the rate of threshold shift can be decreased dramatically. For example, by oscillating bias between +20V and -20V, threshold shifts on the order of only 0.8 V can be demonstrated over time scales as long as 40,000 hours. Such methods have been demonstrated successfully on other technologies, such as liquid-crystal displays. The use of reverse bias can reset the drive transistor, removing all the  $V_{th}$  shift due to forward bias, or slow the degradation of the drive transistor, by periodically removing some of the  $V_{th}$  shift due to forward bias.

[0006] Unfortunately, EL displays, such as OLED, typically perform as a diode, allowing appreciable levels of current to flow and light to be created only when driven in a forward bias. Therefore, known methods use both forward and reverse bias to slow the degradation of a-Si drive TFTs when driving an EL device. These schemes typically involve a first period during which the drive TFT is driven in forward bias and emits light and a second period during which the drive TFT is driven in reverse bias and therefore does not emit light. This means that the EL device is driven with less than 100% of the possible duty cycle.

[0007] For example, Lo et al., in U.S. Pat. No. 7,116,058, teach modulating the reference voltage of the storage capacitor in an active-matrix pixel circuit to reverse-bias the drive transistor between each frame. Sanford et al., in U.S. Pat. No. 6,734,636, teach modulating one of the supply voltages to an AMOLED panel to reverse-bias the drive transistor while storing data that will be subsequently driven. Andry et al., in U.S. Pat. No. 6,872,974, teach varying the voltage and duration of a reverse bias to remove  $V_{th}$  shift, where the duration is between about 1% and 99.9% of a frame time. Tsuchida, in US 2006/0187154 A1, teaches applying reverse bias less often than per-frame, and specifically every predetermined number of frames. Libsch et al., in U.S. Pat. No. 7,167,169, teach a panel configuration using reverse bias within a frame. Howard, in U.S. Pat. No. 6,858,989, teaches applying to each subpixel a reverse bias that depends on the forward bias that was applied to that subpixel.

[0008] In all these schemes, however, since each light-emitting element only emits light when its drive TFT is not reverse biased, the duty cycle of light emission is less than 100%. Therefore, the drive TFT must operate at higher voltage during forward bias to achieve the same luminance it could with 100% duty cycle, which can actually lead to faster TFT degradation. Further, the reduced duty cycle requires the EL device be driven at a higher instantaneous current density, which can reduce the lifetime of the EL device faster than it would have using a traditional forward bias only driving scheme. Additionally, compared to conventional two-transistor, one-capacitor (2T1C) AMOLED backplane designs, these schemes require more complicated external power supplies, additional pixel circuitry or additional signal lines.

[0009] Alternative schemes use reverse bias in a separate phase than light emission. One such scheme is described by Hasumi et al., in "New OLED Pixel Circuit and Driving Method to Suppress Threshold Voltage Shift of a-Si:H TFT," *SID 2006 Digest* paper 46.2, pgs. 1547-1550. Hasumi et al. apply reverse bias when a display is off in order to slow  $V_{th}$  shift. However, they apply reverse bias frequently, for example, for one minute out of every eleven. While such a model can be appropriate for cell phone displays or other displays with intermittent usage, it does not apply to monitor or television applications, or to long-duration portable appli-



one for each color of subpixels. There can alternatively be a separate threshold for each subpixel. Threshold levels can also be set based on the location of a subpixel on a display. Resetting circuitry **23** can, in response to the staleness signals, reset those drive transistors associated with staleness signals indicating they should be reset; this can be accomplished using reverse bias of the drive transistors. Drive transistors which should be reset, and their containing subpixels, will hereinafter be referred to as "stale"; those which should not be reset, and their containing subpixels, will hereinafter be referred to as "fresh." Note that "fresh" does not imply "new"; fresh transistors can have any amount of aging up to the threshold level. Note also that circuitry **21**, **22**, and **23** can all comprise digital logic, analog electronics, microcontrollers and software, programmable logic, or other hardware types known in the art.

**[0031]** Co-pending applications U.S. Ser. No. 11/962,182 by Leon et al. and U.S. Ser. No. 11/766,823 describe methods for reducing visible burn-in due to  $V_{th}$  shift and other aging factors while the display is operating. Consequently, the present invention, used in combination with the above-referenced applications, can allow aging to occur during normal operation and only reset drive transistors after a predetermined operating time, or at one or more times determined by the condition of the display. Where previous methods combined reverse bias with normal operation, the present invention performs reverse bias apart from normal operation. This advantageously provides increased duty cycle and reduced complexity compared to prior art methods.

**[0032]** Referring to FIG. 3, curve **31** shows a representative curve of an aging signal, for example shift in  $V_{th}$  ( $\Delta V_{th}$ , volts), over a typical display lifetime of 50,000 hours. In this example,  $V_{th}$  can shift around 4V in 50,000 hours. Line **33** represents a selected threshold level of 3. Curve **32** shows the result when the transistor is reset whenever the aging signal exceeds the threshold level. In this case, the staleness signal is false when the aging signal is less than or equal to the threshold level, and true when the aging signal is greater than the threshold level. Any transistor with a true staleness signal is reset. In this example, over the lifetime of the panel, reverse bias is used twice, keeping the  $V_{th}$  shift at or below 3V at all times. This reduces by 1V the headroom required in the display drivers, reducing their cost and power dissipation. Note that FIG. 3 shows only one curve using reverse bias. However, as discussed above, each subpixel's drive transistor can be reset when indicated by the staleness signal for that subpixel. Therefore, any time reverse bias is applied; one or more subpixels on the display can be reset. Fresh subpixels, those whose staleness signals do not indicate they should be reset, can be operated so they are not reset with the stale subpixels, as will be described below.

**[0033]** In the example of curve **32**, reverse bias is performed only twice in the lifetime of the display. Reverse bias can be performed while the display is not in use for displaying images, such as at night or other times when the display is off. The present invention therefore does not reduce the duty cycle with which the EL device is driven, so advantageously does not increase the required drive voltage or instantaneous current density.

**[0034]** Referring back to FIG. 2, resetting a drive transistor can take an amount of time dependent on the amount of  $V_{th}$  shift and the conditions of reverse bias applied. For example, since reverse bias can be performed when the display is off, the resetting circuitry **23** can reverse bias each drive transistor

in a time period greater than one frame time. When reverse bias is performed when the display is off, a user's turning on the display can interrupt the reverse bias. The resetting circuitry **23** can include storage circuitry **24** for tracking which subpixels have been interrupted in the middle of a reverse bias cycle and resume reverse bias when the display is turned off. In this way a drive transistor can be completely reset regardless of how long resetting takes. Storage circuitry **24** can store a progress signal representing that a drive transistor should be reverse biased so that resetting circuitry **23** can apply such reverse bias during one or more time periods when the display is not operating. The progress signal for each subpixel can be the staleness signal, or another a yes-or-no value indicating whether the subpixel is stale. It can alternatively be a counter tracking how long reverse bias has been applied to the subpixel. Alternatively, while reverse bias is applied to a drive transistor, a controller can periodically stop reverse bias, measure the aging signal associated with that transistor, and resume reverse bias if the updated staleness signal does not indicate the transistor has been reset.

**[0035]** Signal-production circuitry **21** can employ several methods to provide an aging signal. Co-pending U.S. Ser. No. 11/962,182, by Leon et al., describes a method for measuring the current passing through each of the subpixels. Co-pending U.S. Ser. No. 11/766,823 describes a method for measuring a voltage associated with each drive circuit. Other methods obvious to those skilled in the art can also be employed with the present invention. Referring to FIG. 4, a 2T1C subpixel **16** as known in the art can comprise a drive transistor **103**, select transistor **104**, and EL device **101**, as shown on FIG. 1. It can additionally comprise a gate electrode **43** of drive transistor **103**, a first voltage source **41**, and a second voltage source **42**. These features will be used in discussion of several embodiments of signal-production circuitry.

**[0036]** Comparison circuitry **22** may comprise a comparator, which can compare the aging signal for a subpixel with a threshold level for that subpixel. The output of the comparator can be used as a staleness signal for that subpixel. Note that any comparison to see whether a value is below a threshold is analogous to a comparison to see whether a value is above a threshold. Such comparisons can therefore be employed with the present invention. Although the staleness signal is carrying yes-or-no information, it does not have to be digital; it can be analog, pulse-width modulated, or other forms known in the art. Measurements of the aging signal for each subpixel can be taken, and reverse bias applied, at predetermined intervals, after a predetermined time, or at times calculated based on what is shown on the display. Measurements can also be taken when measurements of a subpixel in the matrix or a representative subpixel indicate one or more subpixels are stale. For an electroluminescent panel including multiple subpixels, an aging signal and a staleness signal can be produced for each subpixel.

**[0037]** Referring to FIG. 5, in one embodiment, as taught in U.S. Ser. No. 11/962,182 by Leon et al., the aging signal can be the current passing through a subpixel, and the staleness signal can indicate that the subpixel current is below a predetermined threshold, or equivalently that the magnitude of the difference between measured current and some reference current is above a predetermined threshold. To this end, each subpixel **16** can include a first voltage source **41** electrically connected to the drive transistor **103** and a second voltage source **42** electrically connected to the electroluminescent device **101**. The drive transistor can have a gate electrode **43**

electrically connected to a select transistor **104**, as shown in FIG. 4. Note that electrical connection can be made through switches, bus lines, conducting transistors, or other devices known in the art. Signal-producing circuitry **21**, as shown in FIG. 2, can include a measuring circuit **51** for measuring the current passing through the first and second voltage sources at different times to provide an aging signal representing variations in the characteristics of the drive transistor and EL device caused by operation of the drive transistor and EL device over time. The aging signal can be the change in current between an initial measurement and a more recent measurement, expressed as a difference or a percentage. The measuring circuit can comprise, for example, a current mirror **511**, current-to-voltage converter **512**, correlated double-sampling unit **513**, and analog-to-digital converter **514**, as taught in U.S. Ser. No. 11/962,182 by Leon et al. The control signal can be compared to a threshold current to produce the staleness signal associated with each subpixel. Note that per Kirchoff's Current Law the measuring circuit can be attached anywhere in the current path through the drive transistor and EL device; for example, it can be attached between first voltage source **41** and drive transistor **103**, or between electroluminescent device **101** and second voltage source **42**. Similarly, the current can be measured through any node or nodes in the current path; for example, the current passing through the drain and source terminals of the drive transistor (**631** and **633** of FIG. 6) can be measured.

**[0038]** Referring to FIG. 6, in another embodiment, in accordance with U.S. Ser. No. 11/766,823, the voltage across a test current sink can be proportional to a voltage associated with a drive circuit, specifically  $V_{th}$  of the drive transistor. This voltage, an aging signal, can be compared to a maximum desired  $V_{th}$  and the result of the comparison be used as a staleness signal. To this end, each subpixel **16** can be a three-transistor, one-capacitor (3T1C) subpixel to provide an aging signal that is a function of the threshold voltage of the subpixel's drive transistor.

**[0039]** Specifically, the subpixel matrix **15** of FIG. 1 can further include a first voltage source **41** and a current sink **62**. The current sink can be electrically connected to a sink voltage source **602**, which can be for example, a second voltage source **42** or ground. Each drive circuit **102** can include three transistors **103**, **104**, **61** as described herein. Each drive transistor **103** can further include a first electrode **631**, which can be a drain terminal, electrically connected to the first voltage source **41**, a second electrode **633**, which can be a source terminal, and a gate electrode **43**, which can be electrically connected to a select transistor **104**. Each electroluminescent device **101** can be electrically connected to the second electrode of the drive transistor, and through a switch **601** to a second voltage source **42**. Switch **601** can be closed for normal operation. It can be opened while measuring the aging signal to eliminate OLED leakage, which might otherwise cause measurement noise. The select transistor can be connected to row line for example **14a** and column line for example **12a**, as shown in FIG. 1, or to the appropriate row and column lines for each subpixel position in subpixel matrix **15**. The subpixel **16** can also include a storage capacitor **640** as known in the art electrically connected to the gate electrode **43** of the drive transistor **103**.

**[0040]** Each subpixel can further include a readout transistor **61** with a first electrode **611** electrically connected to the second electrode of the drive transistor, and a second electrode **613** electrically connected to the current sink **62**. Either

of the first and second electrodes can be either the source or drain of the readout transistor. The gate electrode **43** of the readout transistor can be electrically connected to the gate electrode of select transistor **104**. The signal producing circuitry **21** can further include a test voltage source **64** electrically connected to the gate electrode **43** of the drive transistor, in this case through select transistor **104** as is known in the art. The test voltage source can be the source driver **11** or other circuitry integrated with the source driver **11**, or separate circuitry.

**[0041]** Signal producing circuitry **21** can further include a voltage measurement circuit **63** electrically connected to the second electrode **613** of the readout transistor. In this embodiment, an aging signal that is a function of the threshold voltage of the subpixel's drive transistor can be provided by first setting the test voltage source **64** to a test potential, thus fixing  $V_g$ , the voltage of the gate electrode **43** of drive transistor **103**. Next the current sink **62** can be set to a test current, thus fixing  $I_{ds}$ , as the test current drawn by the sink **62** is forced through the drive transistor **103** from electrode **631** to electrode **633**. The voltage measurement circuit **63** can then be used to measure the voltage at the second electrode **613** of the readout transistor, which is electrically connected to second electrode **633** of the drive transistor, and can thus be at a potential equal to  $V_s$ , to provide the aging signal. Measuring  $V_s$  for a known  $V_g$  allows calculation of  $V_{gs}$ , which, at a given  $I_{ds}$ , fixes a point on the I-V curve of the transistor, allowing  $\Delta V_{th}$  to be determined by comparison with predetermined unaged characteristics of the drive transistor.

**[0042]**  $\Delta V_{th}$  or  $V_s$  can be used as the aging signal; either can represent variations in the characteristics of the drive transistor caused by the operation of the drive transistor over time. A comparator can determine whether  $\Delta V_{th}$  is above a threshold, or whether  $V_s$  is below a threshold, to provide a staleness signal. Note that there can be some potential drop across readout transistor **61**. This and other effects can cause the aging signal not to be perfectly proportional to  $V_{th}$ . The present invention applies in these cases; corrections for such effects can be for example a fixed gain or offset adjustment.

**[0043]** Note that if the EL device is configured so that its cathode is connected to electrode **633**, the typical direction of current flow in the drive transistor will be from electrode **633** to electrode **631**, the opposite of the embodiment described above. The present invention applies to this case; a current source can be substituted for the current sink, and the measurements taken as described above.

**[0044]** A drive transistor can be reset by any of the methods known in the art for reverse bias. One possible method is changing the values of one or more external voltage supplies. Another is applying a negative gate-to-source voltage bias.

**[0045]** Referring back to FIG. 4, in one embodiment the reverse bias can be accomplished by providing each of the drive circuits **102** with first voltage source **41** and second voltage source **42** which during operation have a difference in potential and are the current supply through the associated drive transistor and EL device. In this case the resetting circuitry **23**, as shown in FIG. 2, includes circuitry for changing the potential difference between the first and second voltage sources and applying a voltage on a gate electrode **43** of the drive transistor to cause the transistor to reset. A drive transistor can be reset by adjusting at least one of the voltage sources so that the first and second voltage sources have substantially equal potentials, and adjusting the gate electrode of the drive transistor to a potential which is different

than the potential associated with the adjusted voltage sources. Substantially equal potentials can be defined, for example, as potentials within a selected tolerance (for example 5%) of each other. For example, for an N-channel drive transistor in a typical non-inverted configuration (for example FIG. 4), the gate potential can be less than the potential of the first and second voltage sources, making  $V_{gs}$  negative as  $V_s$  is greater than or equal to the potential of second voltage source 42. Adjusting the first and second voltage sources to have substantially equal potentials advantageously reduces current flow through the EL device during reverse bias, which reduces EL device degradation during the reverse bias phase.

[0046] For an electroluminescent panel including multiple subpixels, stale subpixels can be reverse-biased in this way. However, the fresh subpixels generally share the first and second voltage sources with the stale subpixels. To avoid reverse biasing fresh subpixels, the gates 43 of the fresh drive transistors can be driven to a potential which is substantially the same as the potentials associated with the adjusted first and second voltage sources, which are substantially equal during reverse bias as described above, or to a potential which introduces forward bias on the drive transistor with respect to the potentials of the adjusted voltage sources. Continuing the N-channel example above, the gates of fresh drive transistors can be driven to a potential greater than or equal to the potential of the adjusted voltage sources. Since the voltage sources have substantially equal potentials, no current will flow, and since the gate potential is the same or introduces forward bias, no reverse bias will occur. It can be advantageous to set the gate potential to introduce neither forward nor reverse bias, i.e.  $V_{gs}=0$ .

[0047] Parasitics, current flow through the EL device, AC coupling, and other effects can cause a voltage difference between the source of a drive transistor (for example 633) and the potential of the second voltage source (for example 42). They can also cause a difference between the output of a source driver (for example 11) and the potential applied to the gate electrode of a drive transistor (for example 43). For example, current flow can cause a voltage drop across EL device 101, or AC coupling across select transistor 104 as select line 12a changes state can cause the gate potential to be less than that supplied by the source driver. The gate potentials of fresh and stale drive transistors can be selected to produce the desired bias condition in the presence of these effects. An EL panel can be characterized to determine the magnitude of these effects, and the gate potentials, or potentials supplied by the source drivers, adjusted appropriately.

[0048] The invention has been described in detail with particular reference to certain preferred embodiments thereof, but it will be understood that variations and modifications can be effected within the spirit and scope of the invention. For example, the present invention can apply to any pixel circuit design. The above embodiments are constructed wherein the transistors in the drive circuits are n-channel transistors. It will be understood by those skilled in the art that embodiments wherein the transistors are p-channel transistors, or some combination of n-channel and p-channel, with appropriate well-known modifications to the circuits, can also be useful in this invention. Additionally, the embodiments described show the EL device in a non-inverted (common-cathode) configuration; this invention also applies to inverted (common-anode) configurations.

[0049] The above embodiments are further constructed wherein the transistors in the drive circuits are a-Si transistors. The present invention can apply to any active matrix backplane that is not stable as a function of time. For instance, transistors formed from organic semiconductor materials and zinc oxide are known to vary as a function of time and therefore this same approach can be applied to these transistors.

[0050] The invention has been described in detail with particular reference to certain preferred embodiments thereof, but it will be understood that variations and modifications can be effected within the spirit and scope of the invention.

#### PARTS LIST

[0051]	10 EL panel
[0052]	11 source driver
[0053]	12a column line
[0054]	12b column line
[0055]	12c column line
[0056]	13 gate driver
[0057]	14a row line
[0058]	14b row line
[0059]	14c row line
[0060]	15 subpixel matrix
[0061]	16 subpixel
[0062]	21 signal-production circuitry
[0063]	22 comparison circuitry
[0064]	23 resetting circuitry
[0065]	24 signal-storage circuitry
[0066]	31 curve without reverse bias
[0067]	32 curve with reverse bias
[0068]	33 line
[0069]	41 first voltage source
[0070]	42 second voltage source
[0071]	43 gate electrode
[0072]	51 current-measurement circuitry
[0073]	61 readout transistor
[0074]	62 current sink
[0075]	63 voltage measurement circuit
[0076]	64 test voltage source
[0077]	101 electroluminescent device
[0078]	102 drive circuit
[0079]	103 drive transistor
[0080]	104 select transistor
[0081]	511 current mirror
[0082]	512 current-to-voltage converter
[0083]	513 correlated double-sampling unit
[0084]	514 analog-to-digital converter
[0085]	601 switch
[0086]	602 sink voltage source
[0087]	611 first electrode
[0088]	613 second electrode
[0089]	631 first electrode
[0090]	633 second electrode
[0091]	640 storage capacitor

1. A method for resetting drive transistors associated with subpixels in an electroluminescent display, comprising:

- a) providing an electroluminescent display having a plurality of subpixels, each subpixel including an electroluminescent device and a drive circuit having a drive transistor for providing current through its associated electroluminescent device;
- b) providing a separate aging signal for each subpixel during operation of the electroluminescent display after a predetermined operating time period by responding as

- a function of the current passing through each of the subpixels or as a function of a voltage associated with each drive circuit;
- c) comparing each of the separate aging signals with a corresponding threshold level to produce a separate staleness signal for each subpixel representing whether or not the associated drive transistor should be reset; and
- d) resetting the associated drive transistors in response to staleness signals that indicate such drive transistors should be reset.
2. The method of claim 1 wherein step d) further includes providing a first voltage source and a second voltage source which have a difference in potential and supply current through the associated drive transistor and electroluminescent device during operation; adjusting at least one of the voltage sources so that the first and second voltage sources have substantially equal potentials; and adjusting the gate of the drive transistor to a potential which is different than the potential associated with the adjusted voltage sources.
3. The method of claim 1 wherein step b) further includes measuring the current passing through drain and source terminals of a drive transistor to provide an aging signal.
4. The method of claim 1 wherein step b) further includes providing a test voltage to a gate electrode associated with a drive transistor; forcing a test current through the drive transistor; and measuring the voltage at a source electrode of the drive transistor to provide an aging signal.
5. The method of claim 1 wherein the threshold levels for all subpixels are equal.
6. Apparatus for resetting drive transistors associated with subpixels in an electroluminescent display, comprising:
- a) an array of subpixels, each subpixel including an electroluminescent device and a drive circuit having a drive transistor for providing current through its associated electroluminescent device;
- b) means effective after a predetermined operating time cycle of the electroluminescent display for producing a separate aging signal for each subpixel that is a function of current passing through its associated drive transistor or voltage associated with its associated drive circuit;
- c) means for comparing each of the separate aging signals with a corresponding threshold level to produce a separate staleness signal for each subpixel indicating whether or not its associated drive transistor should be reset; and
- d) means employing reverse bias to reset the drive transistors associated with staleness signals that indicate such drive transistors should be reset.

7. The apparatus of claim 6 wherein each of the drive circuits includes first and second voltage sources which have a difference in potential and supply current through the associated drive transistor and electroluminescent device during operation; and wherein the resetting means further includes means for changing the potential difference between the first and second voltage sources and applying a voltage on a gate electrode of the drive transistor to cause such transistor to reset.

8. The apparatus of claim 6 wherein the resetting means reverse biases each drive transistor in a time period greater than one frame time.

9. The apparatus of claim 8 further including means for storing a progress signal representing that a drive transistor should be reverse biased and applying such reverse bias during one or more time periods when the display is not operating.

10. The apparatus of claim 6 wherein each subpixel further includes a first voltage source electrically connected to the drive transistor and a second voltage source electrically connected to the electroluminescent device, and wherein the signal producing means includes a measuring circuit for measuring the current passing through the first and second voltage sources at different times to provide an aging signal representing variations caused by operation of the drive transistor and electroluminescent device over time.

11. The apparatus of claim 6 wherein:

the subpixel array further includes a first voltage source and a current sink;

each drive transistor further includes a first electrode electrically connected to the first voltage source, a second electrode, and a gate electrode;

each electroluminescent device is electrically connected to the second electrode of the drive transistor;

each subpixel further includes a readout transistor with a first electrode electrically connected to the second electrode of the drive transistor and a second electrode electrically connected to the current sink;

and the signal producing means further includes a test voltage source electrically connected to the gate electrode of the drive transistor and a voltage measurement circuit electrically connected to the second electrode of the readout transistor to provide an aging signal representing variations caused by operation of the drive transistor over time.

\* \* \* \* \*

专利名称(译)	重置电子显示器中的驱动晶体管		
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摘要(译)

一种用于复位与电致发光显示器中的子像素相关联的驱动晶体管的方法，包括提供具有多个子像素的电致发光显示器，每个子像素包括电致发光器件和具有驱动晶体管的驱动电路，所述驱动晶体管用于通过其相关的电致发光器件提供电流；在预定的操作时间段之后，通过响应作为通过每个子像素的电流的函数或者作为与每个驱动电路相关的电压的函数，在电致发光显示器的操作期间为每个子像素提供单独的老化信号；将每个单独的老化信号与相应的阈值电平进行比较，以便为每个子像素产生表示是否应该重置相关驱动晶体管的单独的过时信号；并且响应于指示应该重置这种驱动晶体管的陈旧信号，重置相关的驱动晶体管。

